

Note: The scientific calculator is allowed in all Electronics II exams.

1. (20%) Suppose that the capacitor C_E in Fig.1 is short in the analysis of small-signal equivalent circuit. Derive the expression of the differential-mode output resistance R_o looking from v_o in terms of transistors' small-signal parameters such as $g_{m1}, g_{m2}, \dots, r_{\pi1}, r_{\pi2}, \dots, \beta_1, \beta_2, \dots, r_{o1}, r_{o2}, \dots, r_{o6}$ and the resistances as shown in Fig.1. (about 3-4% for backward pursuing each transistor)

Note: You need to write out complete analysis processes for every steps with sketching necessary small-signal equivalent circuit for explanation)

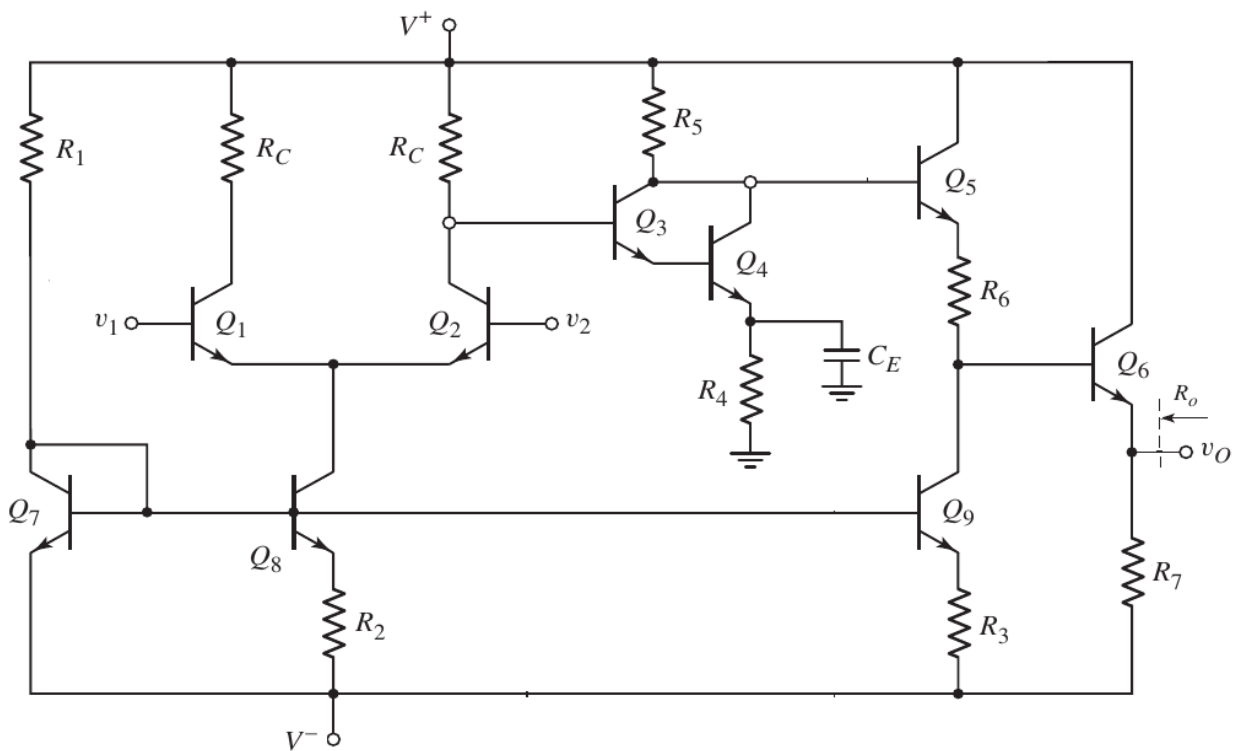


Fig. 1

2. (40%) The transistor parameters for the circuit in Fig. 2 are that

M_1 and M_2 : $K_n = 0.2 \text{ mA/V}^2$, $V_{TN} = 0.8 \text{ V}$, and $\lambda = 0$;

Q_3 and Q_4 : $V_{BE(\text{on})} = 0.7 \text{ V}$, $V_{CE(\text{sat})} = 0.2 \text{ V}$, $\beta = 250$, $V_{A3} = \infty$, and $V_{A4} = 90 \text{ V}$.

Suppose that the output resistance of the current source is $R_o = 200 \text{ k}\Omega$. Design the circuit such that $v_{O2} = 2 \text{ V}$, $I_{C3} = 0.25 \text{ mA}$, and $I_{C4} = 2 \text{ mA}$.

- Determine the maximum value of the common-mode input voltage for v_1 and v_2 . (5%)
- Determine R_1 and R_2 if we make v_o as close to zero as possible. (5%)
- Sketching the small-signal equivalent circuit, derive the expression of the voltage gain v_o / v_{O2} and determine the value. (10%)
- Sketching the small-signal equivalent circuit, derive the expression of the common-mode voltage gain $A_{cm1} = v_{O2} / v_1$. (10%)
- Determine the value of the overall CMRR_{dB} . (5%)
- Determine the output resistance looking from v_o . (5%)

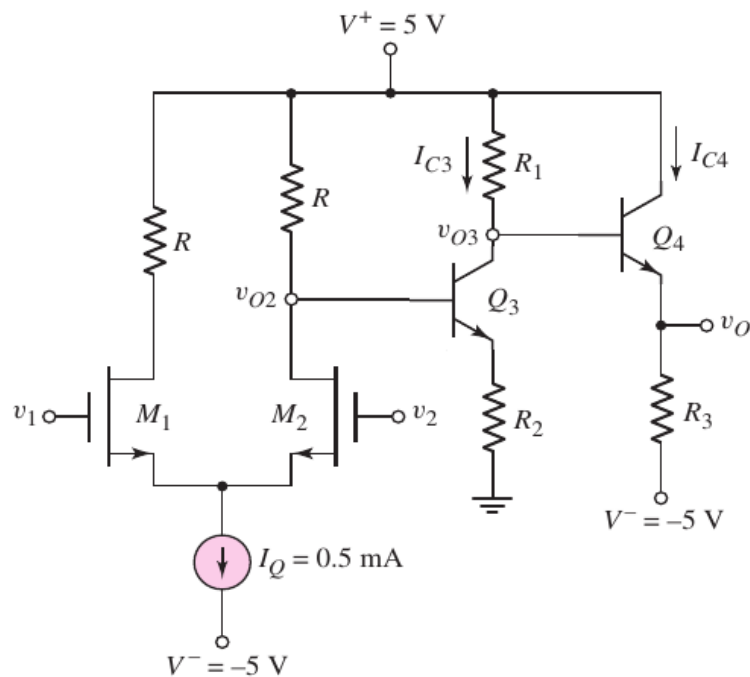


Fig. 2

3. (40%) The differential amplifier in Fig. 3(a) has a pair of NMOS transistors M_3 and M_4 as input devices, a pair of PMOS transistors M_1 and M_2 as an active load in a current mirror configuration, and a pair of NMOS transistors M_5 and M_6 connected as the current source biased with I_Q . Assume that all transistor characteristics are identical, that is, $K_n = K_p$, $\lambda_n = \lambda_p$, and $V_{TP} = -V_{TN}$. Let g_m and r_o denote the transconductance and output resistance of M_1 , respectively. Because the active load circuit is not symmetrical, we can consider the small-signal equivalent half-circuit as split in Fig. 3(b) and let $i_d \approx Gv_1$. Follow the next steps to analyze the differential-mode gain, common-mode gain, and CMRR.
- Sketching the small-signal equivalent circuit, determine the equivalent resistance R_{im} looking into the drain terminal of M_1 . (5%)
 - Considering the case of $v_1 = -v_2 = v_d/2$ for the differential mode, what are R_o , Gv_2 , V_{sg2} , R_{om} , and the differential-mode voltage gain $A_d = v_o/v_d$. (10%)
 - Considering the case of $v_1 = v_2 = v_{cm}$ for the common mode, show that the Norton equivalent circuit model looking into the drain terminal of M_4 is $R_o \approx g_m r_o^2$ and $Gv_2 \approx v_{cm}/r_o$. (10%)
 - Sketching the small-signal equivalent circuit, derive the expression of the common-mode voltage gain $A_{cm} = v_o/v_{cm}$ with the approximation $1 - g_m \left(\frac{1}{g_m} \parallel r_o \right) \approx \frac{1}{g_m r_o}$. (10%)
 - Derive the approximate expression of CMRR in terms of K_n , λ_n , and I_Q . (5%)

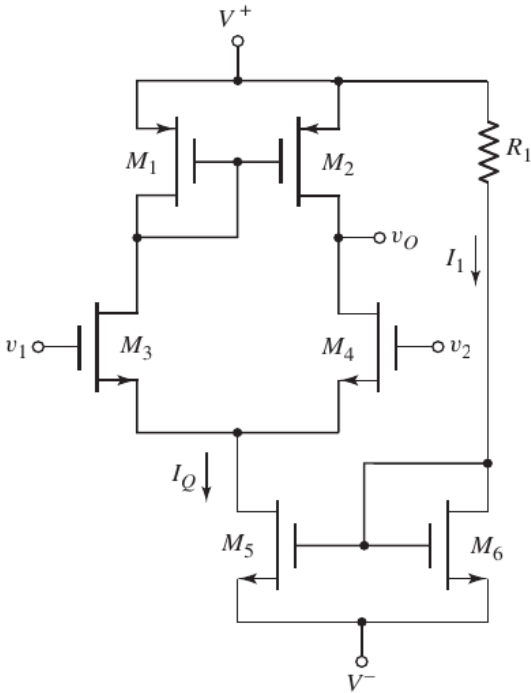


Fig. 3(a)

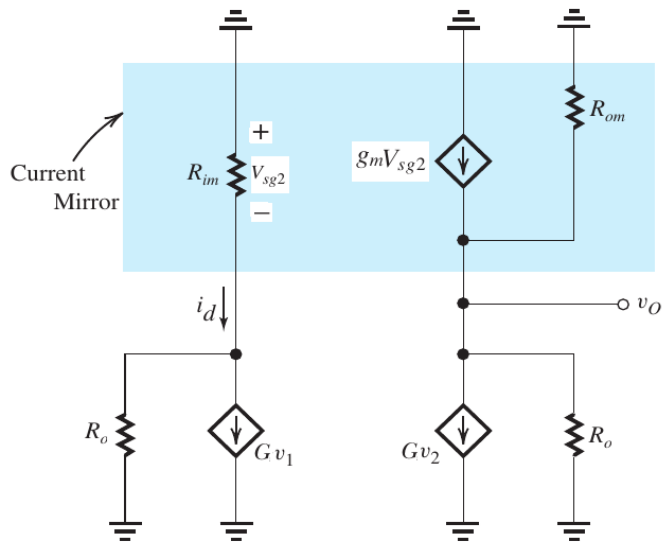


Fig. 3(b)